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## DESCRIPTION

**[0001]** The present invention relates to a method for determining the time of receipt of a radio message, especially an AIS (Automatic Identification System) signal sent from a transmitter to a receiver located in proximity to the transmitter. Furthermore, the invention relates to a receiver, especially an AIS receiver, for receiving such a signal.

**[0002]** For example, in the shipping and aviation fields, the standardized communication system AIS is widely used for local communication between, for instance, moving vessels or between moving vessels and fixed installations, such as a port or a lighthouse. See for example "Recommendation ITU-R M.1371-4", published in 2010 by the International Telecommunication Union.

**[0003]** According to this system, a set of digitally coded, predefined messages of different types can be used by, for example, a vessel for, among other things, informing about its current position and speed to recipients located in the local geographical neighbourhood of the vessel in question. The system is self-organizing in the sense that the participating transmitters divide their respective transmission over a predetermined periodically repeated time window of certain length, so that each transmitter uses a particular part of the time window for transmission.

**[0004]** AIS systems often have low security, in the sense that it is difficult to verify if the contents of the data being sent, such as the position of the transmitter, is accurate and consistent with the real situation. This means that transmitters wishing to participate are expected to follow the AIS standard and to transmit accurate information, and that only simple steps to verify the received information is built into the system. For example, too large timing errors for transmitted signals are not allowed.

**[0005]** It would be desirable to provide an improved control ability of a receiver with respect to the information transmitted by a vessel, such as the vessel position, speed and heading. It would also be desirable to be able to detect whether a particular AIS signal is emitted by one and the same transmitter.

**[0006]** It is known to use directional antennas for measuring the direction to a radio transmitter, in order to verify the transmitter position. This is costly and gives relatively poor precision.

**[0007]** It would be further desirable to provide a simple way for a vessel, whose local clock falls out of sync with the other transmitting vessels, to re-synchronize its local clock. Specifically, it would be desirable to provide a so-called secondary synchronization, which can be used as an alternative to the AIS standard supported synchronization process when the latter for some reason is not available.

**[0008]** It would also be desirable to provide a way for two or more transmitters/receivers to be able to jointly agree on a specific time definition, which is secret among the transmitter/receiver, without any other recipients also receiving information regarding such information.

**[0009]** These objectives should preferably be met using existing AIS standard equipment, in a cost-effective manner and with only a minimum of necessary modifications to existing AIS equipment.

**[0010]** In the field of mobile telephony, US 2007/0276616 A1 discloses a method for accurate time of arrival determination using a correlation between a previously known received signal element with a comparison signal.

**[0011]** The present invention solves the above described problems.

**[0012]** Thus, the invention relates to a method according to claim 1.

**[0013]** In the following, the invention will be described in detail, with reference to exemplary embodiments of the invention and to the accompanying drawings, in which:

Figure 1 shows the structure of a general AIS message;

Figure 2 shows a graph representing a gauss filtered, digitally encoded, exemplifying AIS message;

Figure 3 is an overview diagram illustrating a system according to the invention which can be used in a method according to a first embodiment of the invention;

Figure 4 illustrates the transmitter functionality of an AIS equipment according to the invention;

Figures 5a and 5b illustrate two different variants of the receiver functionality in an AIS equipment according to the invention;

Figure 6 is an overview diagram illustrating a system according to the invention which can be used in a method according to a second embodiment of the invention; and

Figures 7a-7d are respective graphs illustrating a signal processing in accordance with the invention.

**[0014]** AIS is an example of a time-distributed messaging system, under which standardized binary coded radio messages can be sent directly between geographically proximate connected units, without the need for a common central unit via which messages need to be sent. The system is based on AIS-connected transmitters periodically or aperiodically emitting different types of standardized messages, and on that the different transmitters comprised in the system themselves distribute the available bandwidth of each transmission time period

between them.

**[0015]** Figure 3 shows two pieces of equipment 330, 340, permanently installed on land, for receiving and possibly transmitting AIS messages, as well as moving vessels in the form of boats 310, 320 arranged off a shore 300, on the water, which vessels comprise respective equipment for transmitting and receiving AIS messages. AIS messages in the form of radio signals 311, 321 are illustrated schematically.

**[0016]** As is shown in figure 3, at least AIS transmitters 310 and 320 transmit to all AIS receivers that are within the receipt range of the respective AIS transmitter in question. Although all connected devices 310, 320, 330, 340 are not necessarily of the same standardized type, they will, when brought together geographically, establish a self-organizing communication between them based on time division over a certain transmission period. How this works falls within the AIS standard and is conventional as such.

**[0017]** It is understood that the fixed and mobile units 310, 320, 330, 340 illustrated in Figure 3 may equally well be other types of fixed devices, such as a permanently installed transponder at an airport, and other types of moving vessels, such as aircraft or wheeled vehicles.

**[0018]** Different types of AIS messages are intended to use for different types of information, examples comprising information regarding the current position, speed, yaw rate, destination, draught and so on, for the connected device. Different types of AIS messages are transmitted according to the AIS standard at different intervals and using different power, depending inter alia on the type of transmitter.

**[0019]** These and other aspects of the AIS system are controlled, among others, in the recommendation ITU-R M-1371 (above). More information is also present in the documentation provided by the International Maritime Organization (IMO).

**[0020]** Figure 1 illustrates the general data structure for an AIS message, comprising the following binary coded fields. A "bit" is a "one" or a "zero".

- **Preamble:** an initial sequence of 24 bits. May, for instance, be alternatingly "0" and "1", but with at least one trailing "0".
- **Start flag and End flag:** both comprise 8 bits and serve as delimiters for the AIS message. May for instance be 8 bits that are all "1".
- **Data:** 168 bits, apart from the information to be transmitted comprising the message identifier **MSG ID**, the sender identifier **User ID** and the field **Communication state**.
- **FCS:** a 16 bit checksum calculated based upon the contents of the field Data, and the purpose of which is to be able to verify the integrity of the AIS message contents.
- **Buffer:** 24 bits that may be used in predefined ways in order to communicate information about, for instance, sender distance and signal quality.

[0021] Thus, of the total length of 256 bits, a certain part of the field Data is used for the actual transfer of parameter values, while some initial fields, such as Preamble and Start Flag, comprise previously determined information.

[0022] Figure 4 illustrates the functional structure concerning transmission of an AIS device 400 according to a preferred embodiment, comprising means for assembling and periodically transmitting AIS messages of at least one predetermined type, which AIS messages are readable by a receiving device for AIS messages. It is preferable that the transmitting device 400 follows the specifications of a particular AIS type (such as "Class A", "Class B", base stations, AtoN (Aids to Navigation), SART (Search and Rescue Transmitter), and so on), and as such is capable of sending a certain set of different AIS message types at certain intervals.

[0023] The device 400 comprises a sensor 401, arranged to detect that an AIS message of the certain predetermined type, or possibly other types, is to be sent, such as based upon the time frame allotted to the transmission device 400 in the distributed system, or based upon CSTDMA.

[0024] Further, the transmitting device 400 comprises a digital message assembly means 402, arranged to assemble, digitally store and calculate a checksum for an AIS message of said predetermined, and any other, types. The message assembly means 402 receives the information to be transmitted in the AIS message inter alia, or at least, from a data source 403, which can contain information such as geographic position, transmitter identity and so on.

[0025] An example of the message produced and stored by the message assembly means 402 is illustrated in Figure 2, as the sequence of binary "ones" and "zeroes" shown at the bottom of the figure. In Figure 2, the message is abbreviated for reasons of clarity.

[0026] In a preferred embodiment, the transmitting device 400 further comprises a filtering means 404, arranged to produce a filtered signal based on said digitally stored binary digital signal. One example of such a filtered signal is shown in Figure 2, as the upper curve. The binary "ones" and "zeroes" are shown in figure 2 next to the filtered signal, and it is clear from figure 2 that a binary "one" corresponds to a higher signal value in the curve, and vice versa for a binary "zero", which corresponds to a lower signal value in the curve. As illustrated in Figure 2, the message does not follow the format regarding the number of bits in different fields illustrated in figure 1.

[0027] As is clear from figure 2, the filtered curve is not a pure step function, but a filtered version of a step function corresponding to the number sequence shown below the curve in the figure. This filtering causes the characteristic appearance of the curve. Examples of preferred filters comprise filtering using a gauss filter, as in conventional GMSK modulation (Gaussian Minimum Shift Keying). It is preferred that the obtained filtered curve is analogue, whereby the filter preferably can be an analogue filter, but it may also be digitally encoded, in which case the filter is digital.

**[0028]** The filtered curve is fed to a modulating device 405, which modulates, preferably frequency modulates, the filtered signal on a carrier wave, which is then amplified and fed to an antenna 406, which in turn transmits the AIS signal as a radio signal 420.

**[0029]** It is preferred that all steps 401-406 constitute software or hardware implemented parts, respectively, of one and the same computer equipment, which in that case comprises or constitutes the device 400.

**[0030]** Figure 5a schematically illustrates the receiving functionality of an AIS device 500 according to the present invention, arranged to determine the time of receipt by the AIS device 500 of an AIS message sent by a transmitter similar to the one described above. An antenna 501 is arranged to receive an incoming radio signal 520 encoding an AIS message, so that an analogue electrical signal is achieved. The radio signal, which is a modulated radio signal, is composed of a carrier wave on which a data signal is modulated, preferably frequency modulated. The radio signal is a first signal of the present invention.

**[0031]** The received analogue signal constitutes a second signal of the present invention. This second signal is sampled, and possibly also demodulated. The thus produced, sampled and optionally demodulated signal constitutes a third signal of the invention. In case the signal is both sampled and demodulated, the sampling and demodulation may be performed in any order, see below.

**[0032]** According to the present exemplifying embodiment, the second signal is demodulated and sampled, which sampling is performed using an AD converter 503, so that a digitally stored, sampled and demodulated signal is achieved, corresponding to the demodulated analogue signal. The AD converter 503 has a sampling frequency of at least 10 kHz, preferably at least 100 kHz. Alternatively, the sampling frequency is at least 10 times as high as the bit rate per second of the received signal, which bit rate in the case of an AIS message is 9600 Hz.

**[0033]** Figure 5a shows a demodulation device 502, which is arranged to accept said analogue electrical signal and to demodulate this signal so that a demodulated analogue signal is achieved, and is fed to the AD converter 503. This demodulated analogue signal resembles the upper curve in figure 2 as regards its appearance, but also comprises some components in the form of noise and any additional filtering effects from the transmitter as well as the receiver.

**[0034]** Figure 5b is identical to figure 5a, but illustrating an alternative configuration, wherein the AD converter 503 is arranged to sample the incoming frequency modulated, analogue signal from the antenna 501, so that a digital version of this signal is achieved, and wherein the demodulating device 502 is digital, and performs demodulation in the digital domain.

**[0035]** It is also understood that in case the second signal is sampled directly, either a digital demodulation can be performed, alternatively the information can be directly extracted from the sampled signal without any particular demodulation.

**[0036]** The demodulated digital signal is then fed to a decoding device 504, arranged to determine the AIS message data content based on the demodulated signal, as a stream of data bits similar to the lower number sequence in figure 2. This decoding is done in a conventional manner. The decoding device 504 is preferably an AIS standard modem, which modem may also comprise, for example, the demodulating device 502.

**[0037]** The decoding device 504 is arranged to, after decoding, make the received message available via an interface 505, such as a graphic display or a digital communication interface for the publication of the message for reading of external software modules.

**[0038]** According to the invention, a predetermined signal element has been defined beforehand which it is known that said stream of data bits comprises. Preferred examples of such predetermined signal elements are the front edge or the trailing edge of the field "Start flag", illustrated in figure 1 and in figure 2. For example, at the front edge, it may be known in advance that one "zero" is followed by a series of eight "ones". At the trailing flank, it may be known in advance that a series of eight "ones" is followed by one "zero". A signal element of the latter type is indicated by an arrow 201 in figure 2. It is preferred that the predetermined signal element in a corresponding way is constituted by a predetermined position in a number of bits of the AIS message whose data content is known in advance, and which preferably comprises either a series of "zeroes" or a series of "ones" of a pre-known length, which is followed by a change from "0" to "1" or from "1" to "0". This provides a simple and unambiguously identifiable upward or downward flank in the demodulated signal. It is especially preferred, in particular in AIS applications, that the signal element is constituted by the final bit in a previously known series of bits that each is either "0" or "1", and which is followed by a switch to "1" or "0", respectively. In particular, the end of, in other words the trailing edge of, the field "Start flag" in AIS messages is useful.

**[0039]** According to the invention, it is the time of receipt of said signal element that is determined, and the time of receipt of the message is then determined based on the timing of the predetermined signal element by the position of the predetermined signal element in the message being known in advance.

**[0040]** This is accomplished by a digitally stored constructed first comparison signal being generated by a timing determination device 508 and on the basis of the stream of data bits, in other words the data content in the received AIS message, so that the constructed comparison signal is constructed to correspond to either the received frequency modulated signal before or, preferably, after demodulation. This comparison signal must always correspond to the sampled signal, but in case the AD converter 503 is arranged upstream of the demodulation device 502 it may be the sampled signal before or after demodulation that the comparison signal is to be constructed to correspond to. In the case in which the signal received by the antenna, before modulation by the transmitter, is filtered by a particular filter, it is preferred that the comparison signal is constructed by, in a corresponding manner, filtering a digital signal representing the data content using a filter with essentially the same properties as the said



certain filter, for example a gauss filter according to the above.

**[0041]** As is clear from figures 5a and 5b, the decoded binary data in the AIS message is fed from the decoding device 504 to the timing determination device 508. In addition, the sampled signal, or alternatively the sampled and demodulated signal, as well as interrupt signals (see below), are fed to the timing determination device 508.

**[0042]** In the present context, that the comparison signal is "constructed to correspond to" the received signal means that the timing determination device 508 digitally synthesizes a signal curve, by data processing of the digital contents of the received AIS message, such as in terms of binary "ones" and "zeroes", in a way that simulates or corresponds to the signal processing in the transmitter that gave rise to the sent and then received, modulated radio signal 520. In other words, the decoded binary data in the AIS message is used to simulate the sampled and optionally demodulated signal received by the timing determination device 508, using known characteristics of the transmitter regarding how the transmitter creates the transmitted radio signal.

**[0043]** In the example illustrated in figures 4, 5a and 5b, this indicates that a gauss filtered curve of the type illustrated at the top in figure 2 is achieved, corresponding to the bit sequence in the AIS message analysed in the decoding device 504. Thus, in this example, the signal having been received by the antenna 501, before modulation by the transmitter, has been filtered by a particular filter which then also is applied to the constructed signal.

**[0044]** Thereafter, an optimum value for a time variable is determined by the timing determination device 508, which time variable specifies a position or displacement of the constructed comparison signal on the time scale relative to said sampled and possibly demodulated signal, and for which optimum value a correlation between the constructed comparison signal and the sampled signal is maximal. In the following, the sampled and optionally demodulated signal is denoted the "sampled signal", for reasons of simplicity.

**[0045]** Herein, the term "correlation" refers to a comparison operation performed between the two signals for individual points in time, and which is a measure of the total conformity and/or covariance of the signals as seen over a certain time interval.

**[0046]** In other words, the constructed comparison signal is oriented by a certain time scale displacement relative to the corresponding sampled signal, after which a correlation between the signals is calculated, and the displacement that maximizes the correlation is the optimum value for the time variable.

**[0047]** The correlation can for example be calculated as:

$$C_j = k \sum_{i=T_0}^{T_1} (S(i)J(i+j))$$

where

$j$  = the said time constant

$C_j$  = the correlation for time constant =  $j$ ,

$k$  = a constant,

$\{T_0, T_0\}$  = the time interval in question,

$S(i)$  = the sampled signal curve in point (i), and

$J(i)$  = the constructed comparison signal curve in point (i).

**[0048]**  $\{T_0, T_0\}$  may cover the whole time overlap of the signal curves, alternatively only a shorter test time interval.

**[0049]** It is realized that the correlation, in a corresponding manner, can be calculated over a continuous time interval, rather than for discrete time points.

**[0050]** Hence, the value for  $j$  which maximizes  $C_j$  is calculated.

**[0051]** According to a preferred embodiment, the optimum value for the time variable is calculated by calculating the correlation for several values of the time variable, and then searching for the global maximum for the correlation as a function of the time variable. Numerically, this can for example be performed by first determining an initial temporal relative orientation of the two signals, which may be based on empirically derived values for delays in the receiver 500. Then, the correlation may be calculated for a number of values of the time variable shifted forwards or backwards relative to the original time orientation, so that a rough approximation of a correlation function  $C(j)$  is achieved, by means of which a coarse value for the optimum time variable can be determined. Then, the correlation may be calculated for shorter intervals of the time variable in an interval around the said rough value, in order to achieve a more accurate value for the maximum. This can be repeated several times, and finally the correlation derivative can be calculated, and the maximum may be determined as the value of the time variable for which the derivative function intersects 0. This intersection may for example be calculated by means of linear interpolation. All these calculations are done in a microprocessor in the receiver 500, and on the initiative of the timing determination device 508.

**[0052]** Then, the thus determined optimum value for the time variable is used for correcting the determination of the time of receipt of the predetermined signal element. Information regarding the corrected timing is then published by the timing determination device 508 via the interface 505.

**[0053]** According to a preferred embodiment, the decoding device 504 is arranged to detect

the receipt of the above discussed predetermined signal element, and at such a detection to report the receipt to the timing determination device 508. In this case, it may be a time determination of the receipt by the decoding device 504 of the predetermined signal element which is corrected by the above described time variable.

**[0054]** According to a further preferred embodiment, all relevant delays in the receiver are measured before the AIS message is received, comprising the delay in the demodulation device 502, the AD converter 503 and the decoding device 504. Then, those of said delays that affect such timing information which is reached by the timing determination device 508 and that relate to the receipt of the predetermined signal element, and/or those of said delays that occur before a signal indicating that the predetermined signal element has been received reaches the timing determination device 508, are used for adjusting the timing determination of the predetermined signal element in addition to the adjustment using the optimum time variable. Such delays are generally consistent, predictable, and can therefore be determined in advance. Examples comprise delays in the AD converter 503 itself.

**[0055]** According to one preferred embodiment, in addition to a filtering intended to simulate a filter that is applied by the transmitter before modulation, such as the above described gauss filtration, the constructed comparison signal may also be subjected to a filtration using a filter before computing the said correlations, so that deviations of the sampled signal that occur due to the previously known analogue frequency and/or phase response are simulated before sampling using said filter. Such a filter may for instance be constituted by a suitable FIR filter.

**[0056]** The sampled signal to be correlated with the comparison signal may be fed to the timing determination device 508 from the decoding device 504, but it is preferred that it is fed to the timing determination device 508 directly from the demodulation device 502 or the AD converter 503. What is important is that information concerning the time of receipt of the predetermined signal element is made known to the timing determination device 508.

**[0057]** The receiver includes a clock 507 which provides the timing determination device 508 with the current time. According to a preferred embodiment, the AD converter 503 sends out at least one interrupt signal to indicate a certain sampling time. The interrupt signal is detected, directly or via for example the decoding device 504, by the timing determination device 508, and the sending of the interrupt signal is timed by the clock 507. In this case, the above described time of receipt of the predetermined signal element is based on the time-determined emission of the said interrupt signal, by means of the sending time of the interrupt signal being compared to a certain temporal position of the sampled signal. It is preferred that the interrupt signal whose sending is related to the sampled signal is the interrupt signal which is sent when, or in connection to when, the predetermined signal element was received. As described above, various delays of the demodulating device 502, the AD converter 503, the decoding device 504 and so on, that are previously known, may be compensated for in an appropriate way in order to determine which interrupt signal is sent when the predetermined signal element was received.

**[0058]** According to a preferred embodiment, the AD converter 503 sends a plurality of interrupt signals at predetermined time intervals, preferably one interrupt signal for each sampling point. In this case, the sending is time-determined using several such interrupt signals consecutively sent out by the clock 507, after which each one of the points in time for the receipt of such interrupt signals is adjusted using said previously known time intervals, in order to estimate the time of sending of a certain single interrupt signal, such as the interrupt signal at which the predetermined signal element was received. The thus adjusted time determinations are averaged, so that an averaged time of sending of the certain single interrupt signal is achieved. Finally, the determination of the time of receipt of the predetermined signal element is based on said averaged point in time of the sending of the certain single interrupt signal, by relating the certain single interrupt signal to a specific position on the time scale of the sampled signal.

**[0059]** For example, the interrupt signal at which the predetermined signal element was received may constitute the certain single interrupt signal, and a sequence of interrupt signals which is then detected by the timing determination device 508 may be used to calculate a more accurate value of the sending time of the certain interrupt signal as follows:

$$T_{med} = K + \frac{1}{m} \sum_{i=0}^{m-1} (T_{CLK}(i) - \frac{1}{F_s})$$

where

$T_{med}$  = the time averaged value for the receipt of the certain single interrupt signal,

$K$  = an optional adjusting time constant for taking into consideration delays in the receiver 500,

$m$  = the total number of interrupt signals to be used for the calculation of the average value,

$T_{CLK}(i)$  = the time of the clock 507 when the interrupt signal with number  $i$  from the certain interrupt signal is detected by the timing determination device 508, and

$F_s$  = the samplings frequency.

**[0060]**  $m$  is preferably a number of samplings corresponding to at least 10 bits in the message signal, alternatively at least 100 samplings.

**[0061]** Such a method provides a highly accurate value for the receipt, and thus also the sending of a certain single interrupt signal (in this example the one when the predetermined signal element was received). Thereby, a precisely defined time reference for the sampled signal is achieved, wherein the above described time determination using the constructed signal also becomes highly accurate, in the corresponding way.

**[0062]** It is further preferred that the clock 507 comprises a local oscillator, which is periodically synchronized to the time information in a received GPS signal (Global Positioning System) 530,

which is received through an antenna 506. The extraction of a time signal from the GPS signal 530 is conventional as such. The present inventors have discovered that if the local oscillator consists of a per se conventional crystal oscillator, which is less expensive than, say, an atomic clock, and in case the oscillator is periodically synchronized using GPS signal timing, adequate results are achieved with regard to the timing of the received AIS message. It is thus preferred not to use an atomic clock.

**[0063]** It is preferred that the clock oscillator 507 is periodically synchronized using the time information of the GPS signal 530, preferably at least every minute, more preferably at least every ten seconds, more preferably at least every second. It is preferred that the clock 507 is sufficiently precise to always have a sense of time that is correct to the precision of at the most 100 ns for such synchronization periods.

**[0064]** Using a method according to the present invention, the time of receipt of a message comprising a predetermined signal element can be determined with great accuracy. Also, the receipt of single messages can be timed precisely.

**[0065]** In particular, this applies to AIS messages. As an AIS receiver typically already comprises modules 501, 502, 503, 504 and 506, the timing determination device 508 may, at low cost, be added to the existing equipment, for example in the form of a software module arranged to be executed on the same computer that already executes for instance software constituting the decoding device 504. This way, the receipt of up to 2250 messages per minute may be timed within the scope of the existing AIS system and without significant investment in new hardware.

**[0066]** In the following, a number of applications are described in which such accurate timing can be used.

**[0067]** Figure 3 illustrates a first example, in which the position of the vessel 310 is determined by triangulation between the respective times of receipt of the message signal by at least three different receivers, such as AIS receivers 320, 330, 340. The clocks in each of these receivers 320, 330, 340 have been synchronized ahead of time, to one and the same common view of the current time, preferably by means of all three being of the above described GPS-synchronized type, thereby sharing the same time reference.

**[0068]** All three receivers 320, 330, 340 receive the AIS radio signal A, B, C from the transmitter 310, but at different times. Each of the receivers 320, 330, 340 measures the time of receipt of the same predetermined signal element of the received signal, and a location of the transmitter 310 is then determined by triangulation based on the respective location of the three recipients, which is known in advance, in combination with differences in the respective time of receipt of the said signal element. The triangulation calculations themselves may be performed by a central computer 350, which is connected to the receivers 320, 330, 340. Any movable participating vessels 320, such as boats, trucks or helicopters, may report their current position through an as such conventional wireless data link 322.

**[0069]** Using such a process, the present inventors have noted that accuracies of as little as a few hundred meters have been achieved regarding the position determination of an AIS transmitter 310 at an SNR (Signal to Noise Ratio) of 10 dB, which is sufficient to in practice be able to verify the accuracy of, for example, the reported location of the AIS transmitter 310 as reported by the transmitter 310 itself.

**[0070]** According to a preferred embodiment, at least one AIS message being sent by the AIS transmitter 310 is tapped, so that a stated position for the AIS transmitter can be obtained. In addition, AIS information that allows the time of sending of a specific future AIS message to be predicted is preferably tapped, for instance using information about where in the AIS time window a specific AIS message is periodically sent from the AIS transmitter 310 during normal operation. Then the receipt of the certain future AIS message from the AIS transmitter 310 in question is tapped and timed, and the position of the AIS transmitter 310 is triangulated as described above, based on the receipt of the certain AIS message in question.

**[0071]** Thereafter, the calculated position of the AIS transmitter 310, and/or a heading and/or a velocity, as calculated based on several successive calculations performed by the location of the AIS transmitter 310, may be compared with a position and/or a heading and/or a velocity of AIS transmitter 310 as stated in an AIS message sent by the AIS transmitter 310. In the case in which locations and/or headings and/or velocities are pair-wise different by more than a respective predetermined value, an alarm signal is emitted via an alarm device 351, which in figure 3 is illustrated symbolically.

**[0072]** Figure 6 illustrates a further example of an application of the method described above for precise timing of an AIS message, with at least two transmitter/receiver apparatuses 610, 620, such as AIS equipment that can be installed on fixed and/or movable objects, both including a respective GPS synchronized clock according to the above, and both being able to send out radio signals 611, 621, such as AIS messages. In this case, the position of a transmitter 610 relative to a receiver 620 is known in advance by the receiver 620, whereby the transmitter 610 transmits a radio message including a predetermined signal element, such as an AIS message, to the receiver 620. Thereby, the receiver 620 can determine the timing of the receipt of the transmitted signal with high accuracy, using a method according to the invention. After correcting for the signal delay between the transmitter 610 and the receiver 620, based on said known relative distances, a precise timing of the transmitter's 610 sending of the message is achieved by the receiver 20. Thus, a common point in time can be synchronized between the transmitter 610 and receiver 620, based upon the commonly known timing of the sending by the transmitter 610 of the message signal.

**[0073]** Such a time synchronization between two devices 610, 620 can be used, for instance, to agree on a common secret sense of time, without any other nearby devices being able to gain knowledge about such a sense of time. The sense of time can for example be used to agree ahead of time on the timing of a change of carrier frequency for secret information, or the like.

**[0074]** In both of the applications described in connection with Figures 3 and 6, there are additional benefits of adding another layer of authentication based on a fingerprint regarding the transmitter 310, 610.

**[0075]** For example, in Figure 3, the verification of the position, speed and/or heading of the transmitter 310 may be combined with a verification of such fingerprint, so that the combined verification becomes very safe. For example, the fingerprint of one AIS message may be compared with that of a later sent AIS message, and verification can be done by examining whether or not the fingerprints are identical. Alternatively, a detected fingerprint may be compared to a previously known fingerprint, based on knowledge of the type of AIS transmitter equipment or the like. Moreover, it can be verified that the same radio signal is accessed by all three receivers 320, 330, 340.

**[0076]** In Figure 6, a verification of a previously known fingerprint of the transmitter 610 is used to ensure that the transmitter 610 really is the expected one.

**[0077]** Herein, a "fingerprint" refers to a characteristic of the radio signal depending on the characteristics of the transmitter itself, such as a previously known deviation from an ideal signal depending on signal processing in the transmitter. Examples include predictable variations in timing inaccuracies over an AIS period, characteristic pulse shapes at ramp-ups or ramp-downs and the resulting GMSK signature of the sender.

**[0078]** In the following, an embodiment is presented in order to provide a thorough understanding of the invention, in which example a received AIS signal is simulated, and then compared to a constructed signal of the present invention.

**[0079]** First, a bit stream is created, consisting of the fields "Preamble" and "Start flag", above, and the random bits corresponding to the rest of the AIS message. In total, the bit stream comprises 250 bits, which may then form the basis for a simulation of a sending of an AIS message with bit rate 9600 bits/sec.

**[0080]** The digital bit stream is graphically illustrated in the lower curve of figure 7a. The bit stream then passes to a conventional digital gauss filter, resulting in the upper curve of figure 7a.

**[0081]** To simulate the interference of the signal arising under normal operating conditions in reality, the resulting signal is then degraded using two consecutive first-order low-pass filters with cut-off frequencies around 5 kHz, arranged to simulate the transmitter's signal characteristics. Finally, a white noise is added, with the same root mean square as the signal itself. The resulting curve is illustrated in figure 7b.

**[0082]** A simulated sampling, with a sampling rate of 192 kHz, is then carried out, wherein a sampled signal according to the invention is achieved.

**[0083]** A comparison signal is constructed by the upper curve in figure 7a being subjected to the same low pass filter as the simulated signal, but without white noise.

**[0084]** Next, the sampled and the constructed signals are correlated for different values of a time variable that displace them on the time scale relative to each other. The calculated correlation function of the time variable is illustrated in figure 7c.

**[0085]** Figure 7D is an enlargement of the curve 701 shown in figure 7c around the global maximum of the curve. In Figure 7d, both the actual curve 701 and its derivative 702 are shown. The y-axis indicates the value of the derivative. As is evident in figure 7d, the optimum time variable 700 is selected as the value 703 that maximizes the correlation 701, and which is finally calculated by linear interpolation of the derivative function 702.

**[0086]** Such a simulation of real conditions gives the following repeatability regarding the timing for different relative noise levels:

Signal level = Noise level: about 90% fall within  $\pm 5.00 \mu\text{s}$

Signal level = 10 x Noise level: about 90% fall within  $\pm 0.50 \mu\text{s}$

Signal level = 100 x Noise level: about 90% fall within  $\pm 0.05 \mu\text{s}$

**[0087]** Above, preferred embodiments have been described. However, it is apparent to the person skilled in the art that many modifications can be made to the described embodiments without departing from the basic idea of the invention.

**[0088]** For example, the inventive concept is also applicable to other types of radio messages than AIS messages.

**[0089]** Furthermore, other types of previously known or detected fingerprints regarding the transmitter's signal characteristics may be used to increase security.

**[0090]** Thus, the invention is not to be limited to the embodiments described, but can be varied within the scope of the enclosed claims.

## **REFERENCES CITED IN THE DESCRIPTION**

Cited references



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**Patent documents cited in the description**

- US20070276616A1 [0010]

**Non-patent literature cited in the description**

- Recommendation ITU-R M.1371-4International Telecommunication  
Union20100000 [0002]

## PATENTKRAV

1. Fremgangsmåde til bestemmelse af positionen for en sender (310), som udsender en AIS (Automatic Identification System) -meddelelse, **kendetegnet ved, at** i det  
5 mindste tre modtagere (320, 330, 340), hvis respektive taktgivere (507) på forhånd er synkroniserede, modtager den nævnte AIS meddelelse som ét og samme første, binærkodede og modulerede radiosignal (520) fra senderen (310), **at** det første signal modtages af den respektive modtager (320, 330, 340), under anvendelse af en antenne (501), hvilket resulterer i et andet signal i form af et analogt elektrisk signal, **at** hver af  
10 modtagerne bestemmer tidspunktet for modtagelse af det samme forudbestemte signalelement i det første radiosignal (520), **at** en position for senderen bestemmes ved triangulering, baseret på den respektive position af de tre modtagere i kombination med forskelle i de respektive tidspunkter for de respektive modtagelser af det nævnte signalelement, og **at** det respektive tidspunkt for det nævnte signalelement bestemmes,  
15 af hver respektiv enkelt af de nævnte modtagere (320, 330, 340), ved at eksemplere det nævnte andet signal, hvilken eksemplering kan udføres i en vilkårlig rækkefølge i forhold til enhver demodulation af det andet signal, hvilken eksemplering udføres under anvendelse af en AD konverter (503), hvilket resulterer i et digitalt lagret eksempleret signal som et digitalt lagret tredje signal, **at** et dataindhold i meddelelsen bestemmes ud fra  
20 det tredje signal som en strøm af databit, hvilken strøm af databit omfatter det nævnte forudbestemte signalelement, **at** en optimal værdi bestemmes for en tidsvariabel, hvilken tidsvariabel indikerer en tidsposition for et digitalt sammenligningssignal i forhold til det tredje signal, hvilket digitalt sammenligningssignal er konstrueret til at simulere det eksemplerede signal og baseret på det nævnte strøm af databit, for hvilken optimal værdi en korrelation imellem det konstruerede sammenligningssignal og det tredje signal er  
25 maksimal, og **at** den nævnte optimale værdi for tidspositionen anvendes til at korrigere bestemmelsen af modtagelsestidspunktet for det forudbestemte signalelement.

2. Fremgangsmåde ifølge krav 1, **kendetegnet ved, at** det tredje signal opnås ved analog modulation af det andet signal, efterfulgt af eksemplering af det resulterende tilsvarende modulerede signal.

3. Fremgangsmåde ifølge krav 1 eller 2, **kendetegnet ved, at** det første signal, før modulation af senderen (310; 400; 610), filtreres under anvendelse af et bestemt filter, og **at** sammenligningssignalet konstrueres ved, på tilsvarende måde, at filtrere et digitalt  
35

signal, som repræsenterer det nævnte dataindhold, under anvendelse af et filter med essentielt samme egenskaber som det nævnte bestemte filter.

4. Fremgangsmåde ifølge krav 1 eller 2, **kendetegnet ved, at** det konstruerede sammenligningssignal, før beregning af de nævnte korrelationer, filtreres under anvendelse af et filter således, at afvigelser i det tredje signal, som opstår som en konsekvens af det analoge frekvens- og/eller fasesvar for den respektive modtager (320, 330, 340; 500; 620) før eksemplering, simuleres under anvendelse af det nævnte filter.
5. Fremgangsmåde ifølge ethvert af de foregående krav, **kendetegnet ved, at** AD-konverteren (503) udsender i det mindste et interrupt-signal til at markere et bestemt eksempleringstidspunkt, **at** tidspunktet for udsendelse af et sådant interrupt-signal bestemmes under anvendelse af en taktgiver (507) i den respektive modtager (320, 330, 340), og **at** tidsbestemmelsen af modtagelsen af det nævnte forudbestemte signalelement er baseret på den tidsmæssige udsendelse af det nævnte interrupt-signal, ved at relatere udsendelsen af interrupt-signalet med en bestemt tidsposition for det tredje signal.
6. Fremgangsmåde ifølge krav 5, **kendetegnet ved, at** AD-konverteren (503) udsender et antal interrupt-signaler med forudbestemte tidsintervaller, hvoraf tidspunktet for udsendelsen af adskillige sådanne respektive interrupt-signaler bestemmes under anvendelse af den nævnte taktgiver (507), og justeres under anvendelse af det nævnte tidsinterval med henblik på at estimere tidspunktet for udsendelse af et bestemt interrupt-signal, **at** disse justerede tidsbestemmelser midles således, at en midlet tidsbestemmelse for udsendelsen af det bestemte interrupt-signal opnås, **at** tidsbestemmelsen af modtagelsen af det nævnte forudbestemte signalelement er baseret på den midlede tidsbestemmelse af udsendelsen af det bestemte interrupt-signal ved at relatere det bestemte interrupt-signal med en bestemt tidsposition for det eksemplerede signal.
7. Fremgangsmåde ifølge krav 5 eller 6, **kendetegnet ved, at** taktgiveren (507) omfatter en lokal oscillator, som regelmæssigt synkroniseres med tidsinformationen i et modtaget GPS (Global Positioning System) -signal (530).
8. Fremgangsmåde ifølge ethvert af de foregående krav, **kendetegnet ved, at** en bestemmelse af positionen for senderen (310) udføres af en central enhed (350), som kommunikerer med de nævnte i det mindste tre modtagere (320, 330, 340), baseret på

den respektive position af de tre modtagere (320, 330, 340) i kombination med forskelle i det respektive tidspunkt for det respektive modtagelsestidspunkt.



Fig. 3

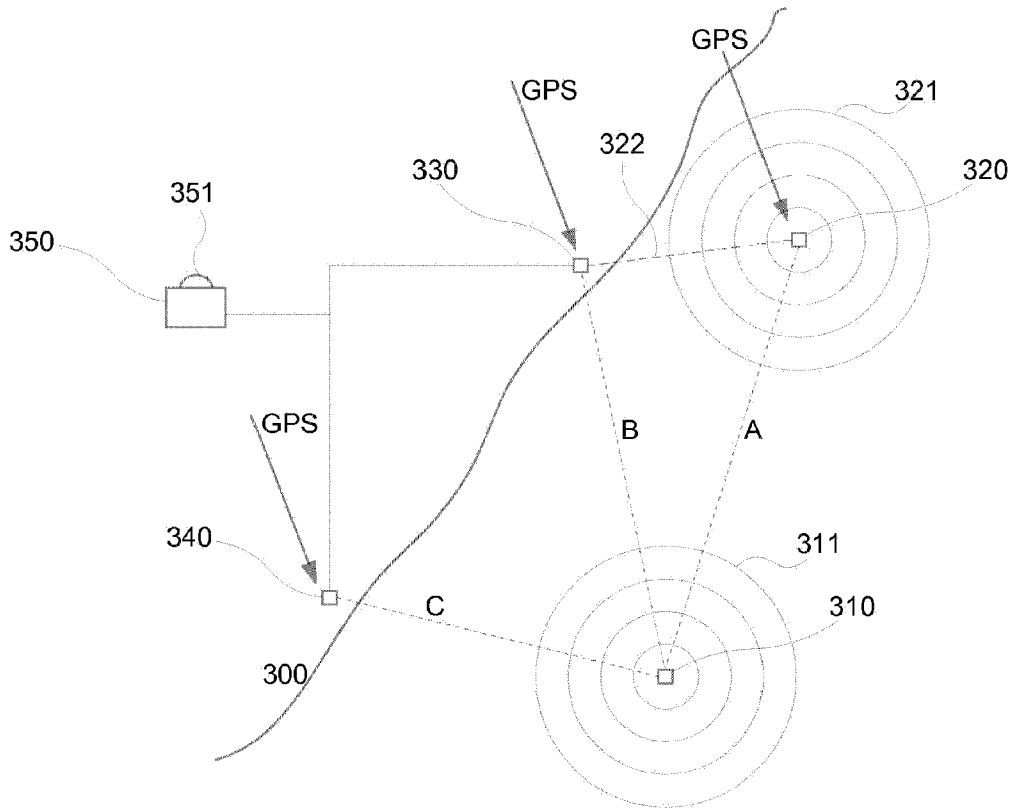


Fig. 4

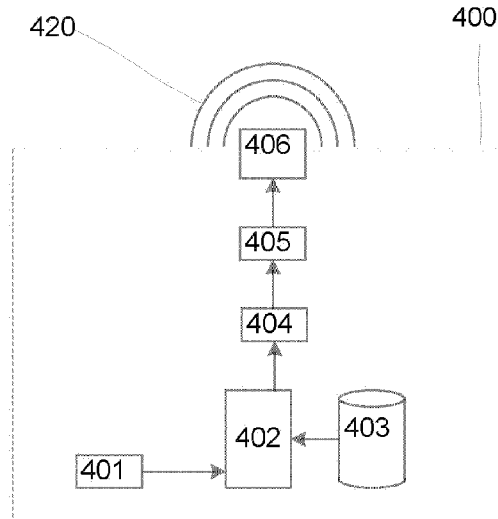


Fig. 5b

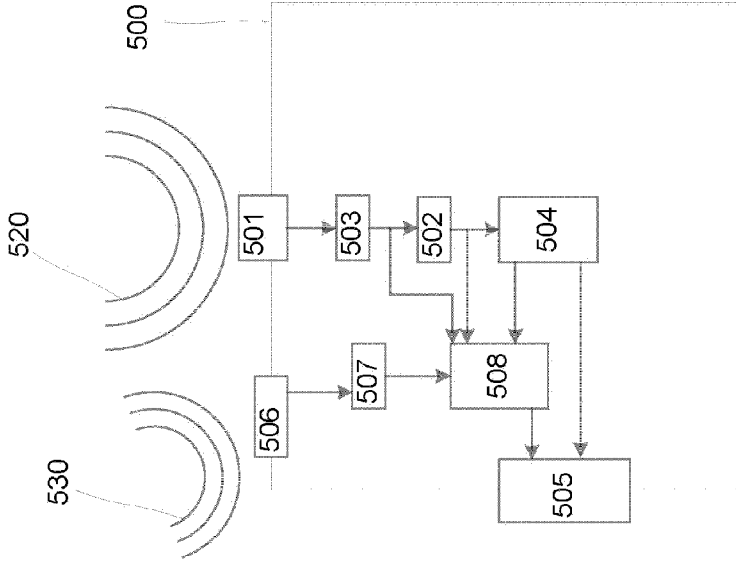


Fig. 5a

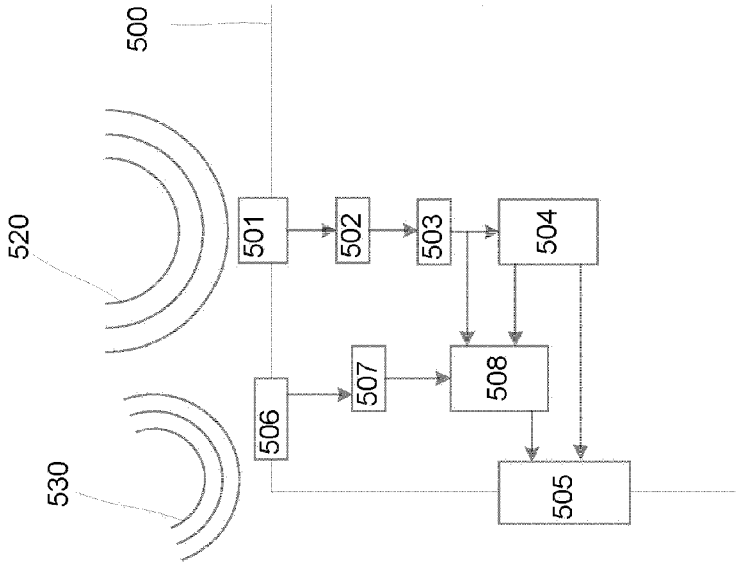


Fig. 6

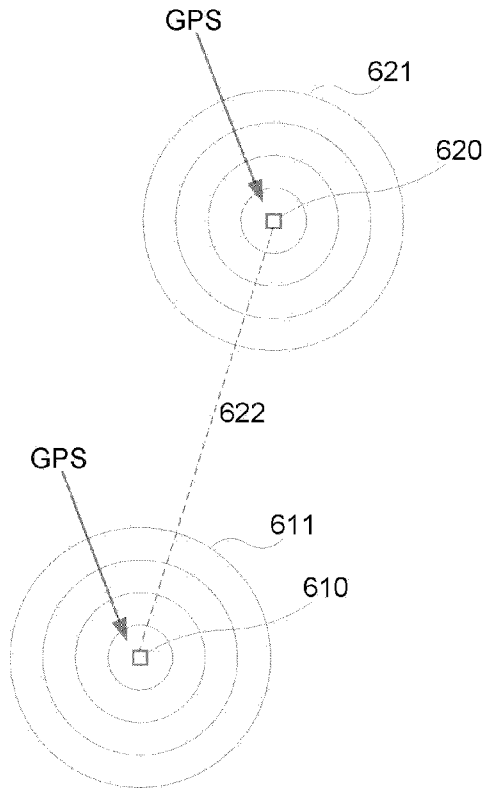




Fig. 7a

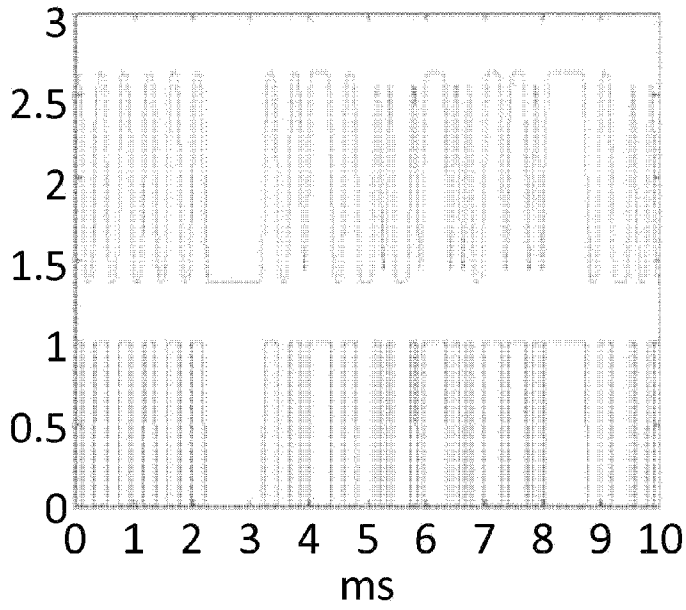


Fig. 7b

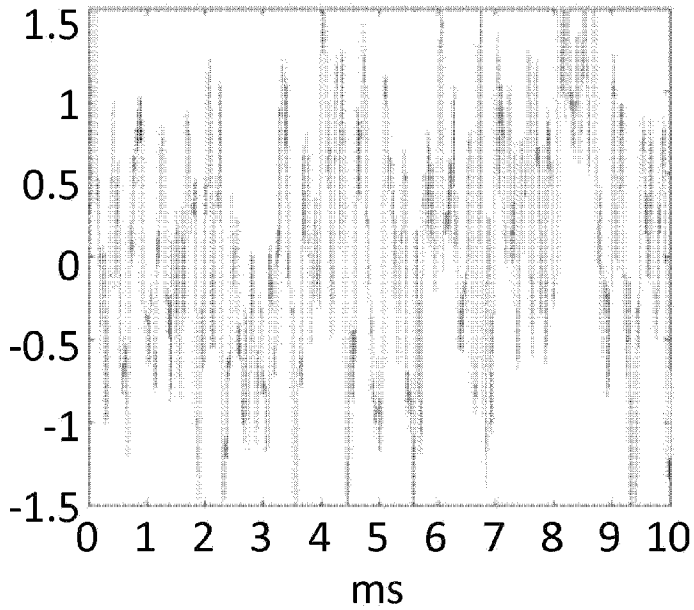


Fig. 7c

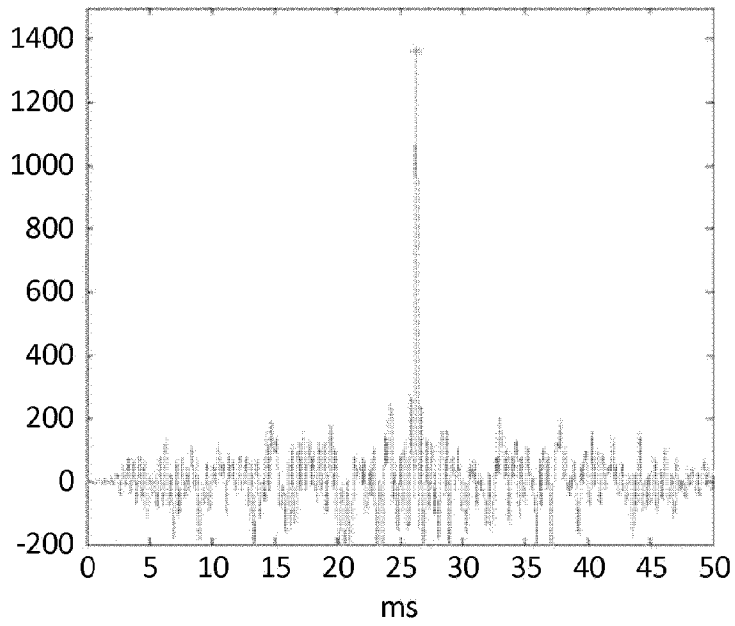


Fig. 7d

